

IN THE CLAIMS

1. (Canceled)

2. (Canceled)

3. (Canceled)

4. (Canceled)

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7. (Canceled)

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14. (Canceled)

15. (Currently amended) A system for converting an input voltage VIN to a digital

output, comprising: K linear flash-type analog-to-digital (A/D) converter apparatuses Z1, Z2, . . . , ZK respectively characterized by reference voltage step sizes $\Delta V_1, \Delta V_2, \dots, \Delta V_K$ and respectively adapted to convert VIN into multibit strings S1, S2, . . . , SK, wherein $\Delta V_1 < \Delta V_2 < \dots < \Delta V_K$, and wherein K is greater than or equal to 2; and encoder means for combining applying a scale factor to at least a selected one of S1, S2, . . . , and SK to generate the digital output, wherein the digital output has a sufficient number of bits to preserve the accuracy that is contained within S1, S2, . . . , and SK.

16. (Previously presented) The system of claim 15, wherein S1, S2, . . . , and SK each have a same number of bits.

17. (Previously presented) The system of claim 15, wherein S1, S2, . . . , and SK do not each have a same number of bits.

18. (Previously presented) The system of claim 15, wherein for k=1, 2, . . . , K the A/D converter apparatus Zk comprises an arithmetic unit Ak in series with an A/D converter Bk, wherein the A/D converters have a same working voltage range, wherein VIN is within the working voltage range, wherein the working voltage range comprises K contiguous voltage subranges denoted as $\delta V_1, \delta V_2, \dots, \delta V_K$ in order of lower to higher voltages, wherein for k=1, 2, . . . , K the arithmetic unit Ak is adapted to change VIN into a new input voltage VIN,k in accordance with a transformation of δV_k into the working voltage range and A/D converter Bk is adapted to transform VIN,k into the multibit string Sk.

19. (Currently amended) The system of claim 18, wherein $\delta V_1, \delta V_2, \dots, \delta V_K$ have values such the relative error of the digital output relative to VIN is a piecewise continuous function of VIN within the working voltage range, said piecewise continuous function of VIN, the piecewise continuous function having K pieces, wherein the relative error within each said piece of the K pieces is a monotonically decreasing function of VIN, and wherein each piece of the K pieces has about a same maximum relative error.

20. (Previously presented) The system of claim 15, wherein K=2, wherein the A/D converter apparatuses Z1 and Z2 comprise A/D converters B1 and B2 having working voltage ranges δ_1 and δ_2 , respectively, such that δ_2 is a subset of δ_1 and δ_2/δ_1 is an integer subject to $\delta_2/\delta_1 > 1$, wherein B1 and B2 are respectively adapted to convert VIN to S1 and S2, and wherein the encoder means is adapted to generate the digital output as S2 if S2 is not within the voltage range δ_1 else the encoder means is adapted to generate the digital output as S1 multiplied by δ_2/δ_1 .

21. (Previously presented) The system of claim 20, wherein $\delta_2/\delta_1 = 2J$, and wherein J is a positive integer.

22. (Currently amended) A method for converting an input voltage VIN to a digital output, comprising: providing K linear flash-type analog-to-digital (A/D) converter apparatuses Z1, Z2, . . . , ZK respectively characterized by reference voltage step sizes $\Delta V_1, \Delta V_2, \dots, \Delta V_K$, wherein $\Delta V_1 < \Delta V_2 < \dots < \Delta V_K$, and wherein K is greater than or equal to 2; converting VIN, by converter apparatuses Z1, Z2, . . . , ZK, into multibit strings S1, S2, . . . , SK, respectively; and combining S1, S2, . . . , and SK to generate the digital output, wherein the digital output has a sufficient number of bits to preserve the accuracy that is contained within S1, S2, . . . , and SK.

23. (Previously presented) The system of claim 22, wherein S1, S2, . . . , and SK each have a same number of bits.

24. (Previously presented) The method of claim 22, wherein S1, S2, . . . , and SK do not each have a same number of bits.

25. (Previously presented) The method of claim 22, wherein for k=1, 2, . . . , K the A/D converter apparatus Zk comprises an arithmetic unit Ak in series with an A/D converter Bk, wherein the A/D converters have a same working voltage range, wherein VIN is within the working voltage range, wherein the working voltage range comprises K contiguous voltage subranges denoted as $\delta V_1, \delta V_2, \dots, \delta V_K$ in order of lower to higher

voltages, said method further comprising: changing VIN by the arithmetic unit Ak for $k=1, 2, \dots, K$, into a new input voltage VIN,k in accordance with a transformation of δV_k into the working voltage range; and transforming VIN,k by the A/D converter B_k , into the multibit string S_k .

26. (Currently amended) The method of claim 25, wherein $\delta V_1, \delta V_2, \dots, \delta V_K$ have values such the relative error of the digital output relative to VIN is a piecewise continuous function of VIN within the working voltage range, said piecewise continuous function of VIN, the piecewise continuous function having K pieces, wherein each two consecutive pieces of the K pieces are discontinuously joined together, wherein the relative error within each said piece of the K pieces is a monotonically decreasing function of VIN, and wherein each piece of the K pieces has about a same maximum relative error.

27. (Previously presented) The method of claim 22, wherein $K=2$, wherein the A/D converter apparatuses Z_1 and Z_2 comprise A/D converters B_1 and B_2 having working voltage ranges Δ_1 and Δ_2 , respectively, such that Δ_2 is a subset of Δ_1 and Δ_2/Δ_1 is an integer subject to $\Delta_2/\Delta_1 > 1$, wherein B_1 and B_2 are respectively adapted to convert VIN to S_1 and S_2 , and wherein said combining includes generating the digital output as essentially S_2 if S_2 is not within the voltage range Δ_1 else said combining includes generating the digital output essentially as S_1 multiplied by Δ_2/Δ_1 .

28. (Previously presented) The method of claim 27, wherein $\Delta_2/\Delta_1 = 2J$, and wherein J is a positive integer.